Boosting Capacity through Interference in 5G

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creating and sharing knowledge for telecommunications



Team to be credited

Master students: Nuno Coelho (ISCTE-IUL) Ricardo Alberto (IST) João Sande Lemos (IST) Francisco Rosário (IST) Pedro Miguel (IST) Filipe Ennes Ferreira Flávio Brás

Post-doc:

Ivo Sousa (IST)

- 1- Context:
 - Massive MIMO
 - Physical Layer Network Coding (PLNC)
 - In-band Full-duplex (IBFD)
 - Non-orthogonal multiple access (NOMA)
- 2-mMIMO+PLNC+IBFD
- **3- Results**



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MIMO channel is a linear transformation



$$\begin{bmatrix} \mathbf{y}_{1} \\ \mathbf{y}_{2} \\ \vdots \\ \mathbf{y}_{N_{R}} \end{bmatrix} = \begin{bmatrix} h_{11} & h_{12} & \cdots & h_{1N_{T}} \\ h_{21} & h_{22} & \cdots & h_{2N_{T}} \\ \vdots & \vdots & \ddots & \vdots \\ h_{N_{R}1} & h_{N_{R}2} & \cdots & h_{N_{R}N_{T}} \end{bmatrix} \begin{bmatrix} \mathbf{x}_{1} \\ \mathbf{x}_{2} \\ \vdots \\ \mathbf{x}_{N_{T}} \end{bmatrix} + \begin{bmatrix} \mathbf{n}_{1} \\ \mathbf{n}_{2} \\ \vdots \\ \mathbf{n}_{N_{R}} \end{bmatrix}$$
$$\hat{\mathbf{x}}_{ML} = \min_{\mathbf{x}} \left\{ \| \mathbf{y} - \mathbf{H} \mathbf{x} \|^{2} \right\}$$

Example: 3 dimensions (3 antennas using PAM).

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In Massive MIMO: Linear Detection Processing suffices!

Zero-forcing (or MMSE): $(\underbrace{\mathbf{H}^{H}\mathbf{H}}_{\mathbf{Z}})^{-1}\mathbf{H}^{H}\mathbf{y} = \mathbf{H}^{\dagger}\mathbf{y}$

Near optimal diversity [1]: $N_R - N_T + 1 \approx N_R$

Low complexity

- But requires inverse ($\mathcal{O}(N_T^3)$!)

Exploit orthogonalization [2]:

- Neumann series
- Matrix inversion lemma



F. Rosário, F. A. Monteiro, A. Rodrigues, "Fast Matrix Inversion Updates for Massive MIMO Detection and Precoding", in IEEE Signal Processing Letters, Vol. 23, No. 1, pp. 75-79, Jan. 2016.

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Has w₁ Wants w₂



0	/	1
Has w ₂	Terminal	2
Wants w ₁	(-
	$\langle \rangle$	/



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The two way relay channel

Physical layer Network Coding



Two time-slots. Can we do better? \Rightarrow Yes: merge both stages \Leftrightarrow <u>in-band full-duplex</u>.







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Full-duplex is changing a long-established idea

"It is generally not possible for radios to receive and transmit on the same frequency band because of the interference that results."

In Wireless Communications, Cambridge University Press, p. 454, 2003,



by Andrea Goldsmith

Two possible transceiver configurations with full-duplex:

Tx Transmitted signal Tx Direct path Nearby Scatters Reflected path Desired signal

Independent Tx and Rx antennas:

Shared antenna:



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Non-orthogonal multiple access (NOMA)





R. Alberto, F. A. Monteiro, "MIMO-NOMA with more than two users", submitted to ICC 2017 - Paris, France

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 k_R , k_A , k_B SI mitigation gains

Results: how many antennas at the relay?



Target: Integer combination of messages

 $\left[\mathbf{D}_{\mathbf{A}}\mathbf{x}_{\mathbf{A}}(n) + \mathbf{D}_{\mathbf{B}}\mathbf{x}_{\mathbf{B}}(n)\right] \operatorname{mod}_{\Lambda_{\mathbf{C}}},$

Processing: Apply zero forcing filters

$$\mathbf{y}_{\mathbf{P}}(n) = \mathbf{H}_{\mathbf{A}\mathbf{R}}^{\dagger} \mathbf{y}_{\mathbf{R}}(n) + \mathbf{H}_{\mathbf{B}\mathbf{R}}^{\dagger} \mathbf{y}_{\mathbf{R}}(n)$$

= $(\mathbf{H}_{\mathbf{A}\mathbf{R}}^{\dagger} \mathbf{H}_{\mathbf{A}\mathbf{R}} \mathbf{x}_{\mathbf{A}}(n) + \mathbf{H}_{\mathbf{B}\mathbf{R}}^{\dagger} \mathbf{H}_{\mathbf{B}\mathbf{R}} \mathbf{x}_{\mathbf{B}}(n)) +$
 $(\mathbf{H}_{\mathbf{B}\mathbf{R}}^{\dagger} \mathbf{H}_{\mathbf{A}\mathbf{R}} \mathbf{x}_{\mathbf{A}}(n) + \mathbf{H}_{\mathbf{A}\mathbf{R}}^{\dagger} \mathbf{H}_{\mathbf{B}\mathbf{R}} \mathbf{x}_{\mathbf{B}}(n)) +$
 $(\mathbf{H}_{\mathbf{A}\mathbf{R}}^{\dagger} + \mathbf{H}_{\mathbf{B}\mathbf{R}}^{\dagger}) (k_{\mathbf{R}} \mathbf{H}_{\mathbf{R}\mathbf{R}} \mathbf{x}_{\mathbf{R}}(n) + \mathbf{n}_{\mathbf{R}}(n))$

$$= \underbrace{\mathbf{D}_{\mathbf{A}}\mathbf{x}_{\mathbf{A}}(n) + \mathbf{D}_{\mathbf{B}}\mathbf{x}_{\mathbf{B}}(n)}_{\mathbf{A}} + \underbrace{\mathbf{D}_{\mathbf{B}}\mathbf{x}_{\mathbf{B}}(n)}_{\mathbf{A}} + \underbrace{\mathbf{D}_{\mathbf{B}}\mathbf{x}_{\mathbf{B}}(n)}_{\mathbf{B}} +$$

desired component

equivalent total noise

 $\tilde{\mathbf{n}}_{\mathbf{R}}^{\star}(n)$

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Recursive Least Squares for a MIMO relay



J. S. Lemos, F. A. Monteiro, I. Sousa, A. Rodrigues, *"Full-Duplex Relaying in MIMO-OFDM Frequency-Selective Channels with Optimal Adaptive Filtering"*, in Proceedigns of GlobalSIP- 3rd IEEE Global Conf. on Signal and Information Processing, Orlando, Florida, Dec. 2015.

Results: how many antennas at the relay? Massive MIMO Effect:

- SER vs equivalent interfering power $(\sigma_{eq}^2 = k + \sigma^2)$
- $N_T = M_T$, varying $M_R \to \infty$

$$k = k_A = k_B = k_R$$
$$\sigma^2 = \sigma^2_{\mathbf{n}_A} = \sigma^2_{\mathbf{n}_B} = \sigma^2_{\mathbf{n}_R}$$



J. S. Lemos, F. A. Monteiro, "Full-duplex massive MIMO with physical layer network coding for the two-way relay channel", in Proc. Of IEEE SAM - The 9th IEEE Sensor Array and Multichannel Signal Processing Workshop, Rio de Janeiro, Brazil, July 2016

Final slide (message to take home)

Clever exploitations of interference offers larger sum-rates by reaching more degrees of freedom

Space domain (mMIMO) , power domain (NOMA), coding domain (PLNC), and full-duplex is possible (IBFD)

-Bidirectional channel with 1 time slot using MaMIMO at the relay is possible;

- The number of antennas at the relay < 300 (for SER of interest).
- -Channel(s) estimation is always crucial to cancel self-interference.



http://iul.pt/~frmo/

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Lattice-based physical-layer network coding

Compute-and-Forward Protocol:

- Relay forwards a linear function of received codewords
- Explore isomorphism between electromagnetic waves and codebook's additive group properties:

Nested Lattice Codes: $\mathcal{L} = \Lambda_F \cap \mathcal{V}_{\Lambda_C} = \{\lambda = [\lambda_F] \text{mod}_{\Lambda_C}, \lambda_F \in \Lambda_F\}$

$$\phi: \mathbb{Z}_3^2 \to \mathcal{L} = \Lambda_{\mathsf{F}} \cap \mathcal{V}_{\Lambda_{\mathsf{C}}} (\in \mathbb{C}),$$

 $S_{\mathcal{A},i}, S_{\mathcal{B},i} \to \mathbf{x}_{\mathsf{A}}, \mathbf{x}_{\mathsf{B}}:$

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Results: w/ self-interference channel estimation error



$\mathbf{H} = \mathbf{ ilde{H}} + \mathcal{E}_{\mathbf{H}}$



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